

Lecture 4

Multiple Representations of Systems

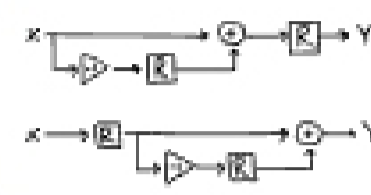
- Operator representations are compact and preserve flow information
 - Combine best features of difference equations and block diagrams
 - Provide a convenient way to reason out a system
- Example: $Y = (1 - R)X$



Operator Expressions Obey Rules of Polynomial Algebra

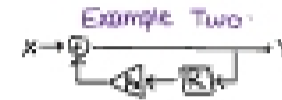
- Operator expressions are commutative and associative, and multiplication distributes over addition
- Equivalent polynomials \rightarrow equivalent systems (when started at rest)

EQUIVALENT REPRESENTATIONS



System Functionals

- We can represent an entire system by its operator representation, which we call the system functional
- Representing a system by a system functional
 - Example one: $Y = HX$
 - Example two: $Y = HX$
- Functionals for systems without feedback are polynomials in R



General Framework for Thinking about Systems

- A system can be represented by a delay or gain or the combination of the two systems using a Cascade or FeedforwardAdd or FeedbackAdd
 - Primitives
 - Delays
 - Gains
 - Composition
 - Cascade
 - FeedforwardAdd
 - FeedbackAdd
- The rules of composition are recursive

Composing Complicated Systems

Any of the rules of composition can be added to any two systems to generate a new type of system



Check yourself: System Functionals

Correct answer: 1

Determine the system functional H



Block's Equation

- Block's equation has two common forms
- Difference between the two: the change of sign G
- The second form is useful in most control applications where the goal is to make Y and X converse



Check yourself: Representing systems

Correct answer: 2

Assume that F and G can be represented by polynomials in R . How many of the following systems can be represented by a polynomial in R .

- Cascade(F, G)
 - $F \times G$ is a polynomial
- FeedforwardAdd(F, G)
 - $F + G$ is a polynomial
- FeedbackAdd($F, \text{Gain}(1)$)
 - $F / (1 - F)$ is generally not a polynomial
- FeedbackAdd($\text{Gain}(1), G$)
 - $F / (1 - FG)$
- FeedbackAdd(F, G)

Check yourself: Representing systems

Correct answer: 5

Assume that each of F and G can be represented by a ratio of two polynomials in R . How many of the following systems can be represented by a ratio of two polynomials?

- Cascade(F, G)
- FeedforwardAdd(F, G)
- FeedbackAdd($F, \text{Gain}(1)$)
- FeedbackAdd($\text{Gain}(1), G$)
- FeedbackAdd(F, G)

System Functionals as Abstraction

Systems that can be represented by a Delay or a gain or by the combination of two systems using Cascade, FeedforwardAdd, or FeedbackAdd can also be represented by a ratio of polynomials in R .

$R = 1/K$ and $K = 1/R$.

Cascade: $\left(\frac{N_1}{D_1} \cdot \frac{N_2}{D_2}\right) = \frac{N_1 N_2}{D_1 D_2}$ Feedforward: $\left(\frac{N_1}{D_1} + \frac{N_2}{D_2}\right) = \frac{N_1 D_2 + N_2 D_1}{D_1 D_2}$ Feedback: $\left(\frac{N_1}{D_1} + \frac{N_2}{D_2}\right) = \frac{N_1 D_2}{D_1 D_2 - N_1 N_2}$

System Functionals can be used interchangeably with primitives

PCAP Framework for Managing Complexity

A system can be represented by a delay or a gain or the combination of two systems using Cascade, FeedforwardAdd, or FeedbackAdd

- Primitives
 - Delays — R
 - Gains — K
- Composition
 - Cascade — FG
 - FeedforwardAdd — $F+G$
 - FeedbackAdd — $F/(1-FG)$
- Abstraction
 - System Functional — ratio of polynomials in R

Poles

Ways to find the poles of a system

- Write the system functional in special form
 - Write the operator expression in special form
 - Substitute $R = 1/z$ and find the roots of the denominator
- All the methods give the same result

Complex Poles

What if a pole has a non-zero imaginary part?

Example: $\frac{Y}{X} = \frac{1}{1-R+R^2} = \frac{1}{1-\frac{R}{2}-\frac{\sqrt{3}}{2}R}$ $= \frac{1}{z^2-z+1}$

The poles are at $z = \frac{1}{2} \pm j\frac{\sqrt{3}}{2}$; corresponding modes are $(\frac{1}{2} \pm j\frac{\sqrt{3}}{2})^n$
 Powers of complex numbers are easy to compute using polar forms
 Express the pole at $z = a+jb$ as $re^{j\theta}$ where $r^2 = a^2 + b^2$ and $\tan\theta = b/a$
 Then the mode is $(re^{j\theta})^n = r^n e^{jn\theta}$
 geometric growth of magnitude
 linear growth of angle

Complex Roots

An isolated complex root can result only from a difference equation with complex-valued coefficients

Example: $\frac{Y}{X} = \frac{1}{1-re^{j\theta}R}$ Corresponding difference equation: $y[n] - re^{j\theta}y[n-1] = x[n]$

Difference equations that represent physical systems have real valued coefficients
 Difference equations with real-valued coefficients generate real-valued outputs from real-valued inputs
 But systems with real valued coefficients can have complex-valued poles
 The sum of the modes associate with complex conjugates is real

Check yourself: Complex Poles

- Correct answer: 2
- Output of a system with poles at $z = re^{\pm j\theta}$
- What are the values for r and θ ? $0.5 < r < 1$ and $\theta \approx 0.5$

Designing a Control System

We have already built several feedback systems to control the robot:

- wallFinder—move forward or backward to position robot a desired distance from the wall in front of it
- wallFollower—move the robot parallel to a wall maintaining a desired distance from the wall

We can use the Signals and Systems Abstraction to gain insight into the general problem of designing a feedback controller

Controlling Accumulation

Both wallFinder and wallFollower contained accumulators to model integrative processes

- wallFinder: command forward velocity to set forward distance
- wallFollower: command rotational velocity to set lateral distance

Integrator: $V \longrightarrow \int dt \longrightarrow D$

Accumulator model: $d[n+1] = d[n] + Tv[n]$



Finding the Optimal Gain

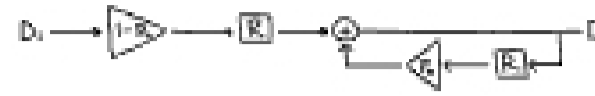
Consider a proportional controller for the accumulator system
 Replace accumulator with equivalent system functional
 This gives us an equivalent system with a single block

Unit Sample Response

The system functional contains a single pole at $p_0 = 1 - KT$

$$\frac{D}{D_i} = \frac{KTR}{1 - (1-KT)R} = \frac{(1-p_0)R}{1-p_0R}$$

The numerator is just a gain and a delay



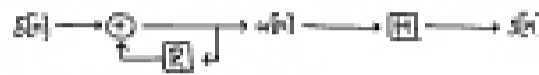
Step Response

We are often interested in the step response of a control system

Start with the output $d[n] = 0$ while the input $di[n]$ is held constant where $d[n]$ is the sensorDistance and $di[n]$ is the desiredDistance

Calculating unit-step response

Unit Step response $s[n]$ is the response of H to the unit-step signal $u[n]$, which is constructed by accumulation of the unit sample $\delta[n]$



The unit step response $s[n]$ is equal to the accumulated responses to the unit sample response $h[n]$

Root Locus

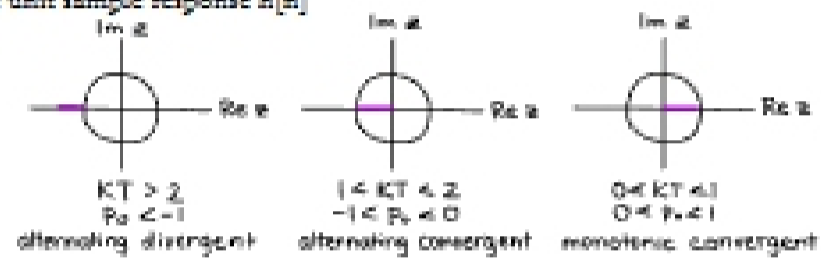
The poles of the system functional provide insight for choosing K

$$\frac{D}{D_i} = \frac{KTR}{1 - (1-KT)R} = \frac{(1-p_0)R}{1-p_0R} \rightarrow p_0 = 1 - KT$$

Check yourself: Root Locus

Correct answer: 2

Find KT for fastest convergence of unit-sample response. $KT = 1$



Destabilizing Effect of Delay

The optimum gain allows the proportional controller system to converge in a single step



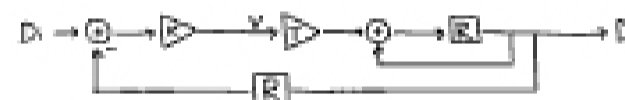
Adding a single delay causes stability



Check yourself: System Functional

Correct answer: 4

Find the system functional: $H = \frac{KTR}{1 - R + KTRz^{-1}}$

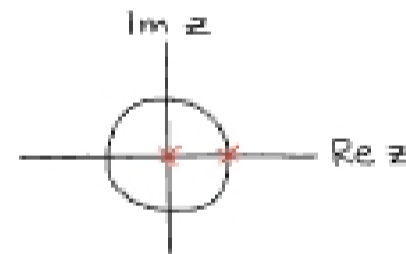


Feedback and Control: Poles

If KT is small, the poles at $z \approx KT$ and $z \approx 1 - KT$

$$z = \frac{1}{2} \pm \sqrt{\left(\frac{1}{2}\right)^2 - KT} = \frac{1}{2} (1 \pm \sqrt{1 - 4KT}) \approx \frac{1}{2} (1 \pm (1 - 2KT)) = 1 - KT, KT$$

Pole near 0 generates fast response
 Pole near 1 generates slow response
 The slow mode dominates the response



Check yourself: Poles

Correct answer: 5

The period of oscillation is 6

Check yourself: Poles

Correct answer: 2

The fastest response is when $KT = 1/4$

Destabilizing Effect of Delay

Adding delay tends to destabilize control systems

Check yourself: Possible Test Question

Correct answer:

How many of the following statements are true?

- This system has three poles
- Unit-sample response is the sum of three geometric sequences
- Unit-sample response is $y[n] = 0, 0, 0, 1, 0, 0, 1, 0, 0, 1, 0, 0, 1, \dots$
- Unit-sample response is $y[n] = 1, 0, 0, 1, 0, 0, 1, 0, 0, 1, 0, 0, 1, \dots$
- One of the poles is at $z = 1$



Designing Control Systems: Summary

System functionals provide a convenient summary of information that is important for designing control systems