

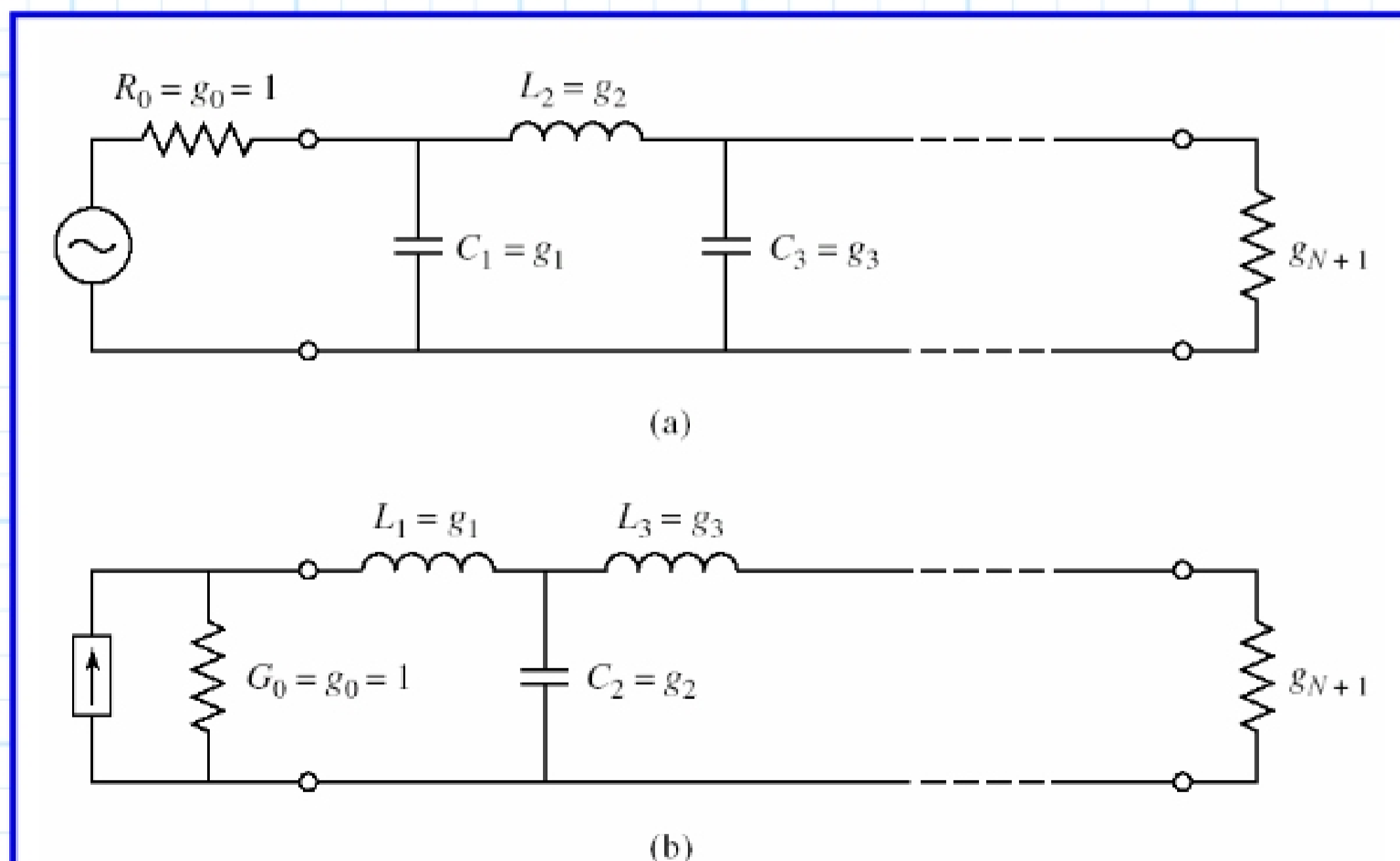
# Filter Realizations Using Lumped Elements

Our **first** filter circuit will be “**realized**” with lumped elements.

**Lumped elements**—we mean inductors  $L$  and capacitors  $C$ !

Since each of these elements are (ideally) perfectly **reactive**, the resulting filter will be **lossless** (ideally).

We will first consider two configurations of a **ladder circuit**:



**Figure 8.25 (p. 393)**

Ladder circuits for low-pass filter prototypes and their element definitions. (a) Prototype beginning with a shunt element. (b) Prototype beginning with a series element.

Note that these two structures provide a **low-pass** filter response (evaluate the circuits at  $\omega = 0$  and  $\omega = \infty$ !).

Moreover, these structures have  $N$  different **reactive elements** (i.e.,  $N$  degrees of design freedom) and thus can be used to realize an  **$N$ -order** power loss ratio.

For example, consider the **Butterworth** power loss ratio function:

$$P_{LR}(\omega) = 1 + \left( \frac{\omega}{\omega_c} \right)^{2N}$$

Recall this is a **low-pass** function, as  $P_{LR} = 1$  at  $\omega = 0$ , and  $P_{LR} = \infty$  at  $\omega = \infty$ . Note also that at  $\omega = \omega_c$ :

$$P_{LR}(\omega = \omega_c) = 1 + \left( \frac{\omega_c}{\omega_c} \right)^{2N} = 1 + 1^{2N} = 2$$

Meaning that:

$$\Gamma(\omega = \omega_c) = \mathbf{T}(\omega = \omega_c) = \frac{1}{2}$$

In other words,  $\omega_c$  defines the 3dB bandwidth of the low-pass filter.

Likewise, we find that this Butterworth function is **maximally flat** at  $\omega = 0$ :

$$P_{LR}(\omega = 0) = 1 + \left( \frac{0}{\omega_c} \right)^{2N} = 1$$

and:

$$\left. \frac{d^n P_{LR}(\omega)}{d\omega^n} \right|_{\omega=0} = 0 \quad \text{for all } n$$

Now, we can determine the function  $P_{LR}(\omega)$  for a lumped element ladder circuit of  $N$  elements using our knowledge of **complex circuit theory**.

Then, we can **equate** the resulting polynomial to the **maximally flat** function above. In this manner, we can determine the appropriate **values** of all inductors  $L$  and capacitors  $C$ !

An **example** of this method is given on pages 392 and 393 of your book. In this case, the filter is very **simple**—just **one** inductor and **one** capacitor. However, as the book shows, finding the solution requires quite a bit **complex algebra**!

Fortunately, your book likewise provides **tables** of complete Butterworth and Chebychev Low-Pass **solutions** (up to order 10) for the ladder circuits of figure 8.25—**no complex algebra** required!